

1.1 V_{OUT}≻ 10 -PT6721 1u⊦ R1 > LOAD Cou PWR GOOD GND GND STBY* REMOTE SENSE (-) = Required 1000µF electrolytic (See footnotes) Cin C_{out} = Required 330µF electrolytic (See footnotes) L1 = Optional 1µH input choke (See formation)

= Optional 1µH input choke (See footnotes)

= Required $10k\Omega$ pull-up when using Pwr Good signal. Pwr good R1

output is high when the output voltage is within specification.

Specifications

Characteristics			PT6721 SERIES			
(T _a = 25°C unless noted)	Symbols	Conditions	Min	Тур	Max	Units
Output Current	Io	$T_a = +60^{\circ}C$, 200 LFM, pkg N $T_a = +25^{\circ}C$, natural convection	$\begin{array}{c} 0.1 \ (1) \\ 0.1 \ (1) \end{array}$	_	14 14	А
Input Voltage Range	V_{in}	$0.1A \le I_o \le 14A$	10.8	_	13.2	V
Output Voltage Tolerance	ΔV_{o}	$\begin{array}{l} V_{in} = +12V, I_o = 14A \\ -40^\circ C \leq T_a \leq +85^\circ C \end{array}$	Vo-0.03	_	Vo+0.03	\mathbf{V}
Short-Circuit Threshold	I _{sc}	$V_{in} = +12V$	_	20	30	А
Line Regulation	Reg _{line}	$10.8V \le V_{in} \le 13.2V$, $I_o = 14A$	_	±5	_	mV
Load Regulation	Reg _{load}	$V_{in} = +12V, 0.1 \le I_o \le 14A$	_	±10	_	mV
Vo Ripple/Noise	V_n	$V_{in} = +12V, I_o = 14A$	_	35	_	mV
Transient Response with C _{out} = 330µF	${f t_{tr}} {f V_{os}}$	I _o step between 7A and 14A V _o over/undershoot	_	50 70	_	μSec mV
Efficiency	η	$\label{eq:Vin} \begin{split} V_{in} = +12V, \ I_o = 9A & V_o = 3.3V \\ V_o = 2.5V \\ V_o = 1.8V \\ V_o = 1.5V \end{split}$	 	90 88 83 81	 	%
Switching Frequency	$f_{ m o}$	$\begin{array}{l} 10.8\mathrm{V} \leq \mathrm{V_{in}} \leq 13.2\mathrm{V} \\ 0.1\mathrm{A} \leq \mathrm{I_o} \leq 14.0\mathrm{A} \end{array}$	300	350	400	kHz
Absolute Maximum Operating Temperature Range	Та	—	-40 (2)		+85 (3)	°C
Storage Temperature	Ts	—	-40	_	+125	°C
Mechanical Shock		Per Mil-STD-883D, Method 2002.3 1 msec, Half Sine, mounted to a fixture	_	500	_	G's
Mechanical Vibration		Per Mil-STD-883D, Method 2007.2, 20-2000 Hz, Soldered in a PC board	_	15	_	G's
Weight	_		_	26	_	grams

Notes: (1) ISR-will operate down to no load with reduced specifications.

(2) For operation below 0°C, Cin and Cout must have stable characteristics. Use either low ESR tantalum or Oscon® capacitors.

- (3) See Safe Operating Area curves, or contact the factory for appropriate derating.
- (4) If the Remote Sense Ground is note used, pin 13 must be connected to pin 14 for optimum output voltage accuracy.

External Capacitors: The PT6721 requires a minimum output capacitance of 330μ F, with a maximum ESR of $50m\Omega @100k$ Hz for proper operation. The maximum allowable output capacitance is $15,000\mu$ F. The input capacitance must be rated for a minimum of 1.6Arms of ripple current. For transient or dynamic load applications, additional capacitance may be required. For more information refer to the application note regarding capacitor selection for this product.

Input Filter: An input filter inductor is optional for most applications. The inductor must be sized to handle 5ADC with a typical value of 1µH.

For technical support and more information, see inside back cover or visit www.ti.com/powertrends



3.3V

5V

5V

12V

12V

VID

Resistor

Resistor

VID

Resistor

PT6702

PT6705

PT6715

PT6721

PT6725

1

1

PT6721—12V

14 Amp Programmable Integrated Switching Regulator

Pin-Out Information

Pin	Function	Pin	Function
1	OVP Drive	13	Remote Sense Gnd (4)
2	Pwr Good	14	GND
3	VID0	15	GND
4	VID1	16	GND
5	VID2	17	GND
6	VID3	18	GND
7	VID4	19	Vout
8	STBY#	20	V _{out}
9	Do not connect	21	Vout
10	Vin	22	V _{out}
11	Vin	23	Remote Sense $\mathrm{V}_{\mathrm{out}}$
12	Vin	#For ST open ground	

12.0V Input Voltage

(Note A)

Programming Information

VID3	VID2	VID1	VIDO	VID4=1 Vout	VID4=0 Vout
1	1	1	1	2.0V	1.30V
1	1	1	0	2.1V	1.35V
1	1	0	1	2.2V	1.40V
1	1	0	0	2.3V	1.45V
1	0	1	1	2.4V	1.50V
1	0	1	0	2.5V	1.55V
1	0	0	1	2.6V	1.60V
1	0	0	0	2.7V	1.65V
0	1	1	1	2.8V	1.70V
0	1	1	0	2.9V	1.75V
0	1	0	1	3.0V	1.80V
0	1	0	0	3.1V	1.85V
0	0	1	1	3.2V	1.90V
0	0	1	0	3.3V	1.95V
0	0	0	1	3.4V	2.00V
0	0	0	0	3.5V	2.05V

Logic 0 = Pin 13 potential (remote sense gnd) Logic 1 = Open circuit (no pull-up resistors)

VID3 and VID4 may not be changed while the unit is operating.

Ordering Information

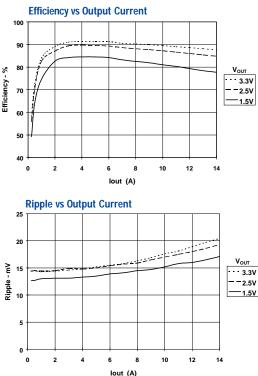
PT6721 = 1.3 to 3.5 Volts (For dimensions and PC board layout, see Package Styles 1300 and 1310.)

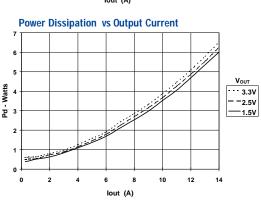
PT Series Suffix (PT1234X)

Case/Pin Configuration

Vertical Through-Hole	N
Horizontal Through-Hole	Α
Horizontal Surface Mount	С

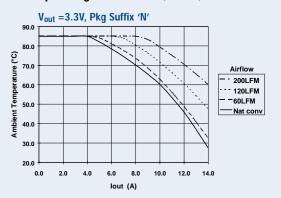
TYPICAL C H A R A C T E R I S T I C S





Note: A All data listed in the above graphs has been developed from actual products tested at 25°C. This data is considered typical data for the ISR. Note: B SOA Curves represent the conditions at which internal components are at or below manufacturer's maximum operating temperatures.

Safe Operating Area Curves (Note B)



🤴 Texas Instruments

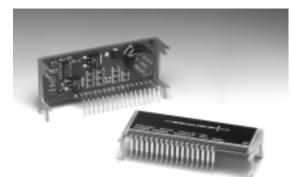
PT6725 Series

13 Amp 12V Input Adjustable Integrated Switching Regulator



SLTS102

(Revised 6/30/2000)

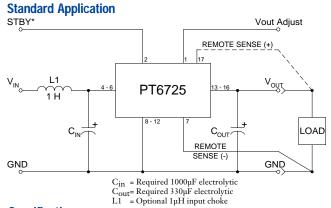


The PT6725 is a new series of high-performance, 13 Amp Integrated Switching Regulators (ISRs) housed in a unique, space-saving 17-pin SIP package. The PT6725 series operates from a standard 12V power bus to provide a high performance lowvoltage power source for the industy's latest high-speed, low-voltage µPs and bus drivers.

EXCALIBUR

The PT6725 series has a trimmed output voltage which may be adjusted with an external resistor. The output current is rated to 14A for output voltages of 2.5V and lower.

Other features includes a differential remote sense which allows the regulator to compensate for voltage drop between the ISR and load. This product also has short circuit protection.



Pin-Out Information

Pin	Function	Pin	Function
1	V _{out} adjust	10	GND
2	STBY*	11	GND
3	Do not connect	12	GND
4	Vin	13	Vout
5	Vin	14	V _{out}
6	Vin	15	Vout
7	Remote Sense Gnd (4)	16	V _{out}
8	GND	17	Remote Sense $\mathrm{V}_{\mathrm{out}}$
9	GND	For S	TBY* pin
			= output enabled

ground = output disabled.

Specifications

Characteristics			P	T6725 SERIE	s	
(T _a = 25°C unless noted)	Symbols	Conditions	Min	Тур	Max	Units
Output Current	I _{o(max)}	$\begin{array}{ll} T_a = +60^{\circ}\text{C}, \ 200 \ \text{LFM}, \ \text{pkg N} & V_o \leq 2.5\text{V} \\ T_a = +25^{\circ}\text{C}, \ \text{natural convection} & V_o > 2.5\text{V} \end{array}$	$\begin{array}{c} 0.1 \ (1) \\ 0.1 \ (1) \end{array}$	_	14.0 (3) 13.0 (3)	А
Input Voltage Range	Vin	$0.1A \le I_o \le 13.0A$	10.8	_	13.2	V
Output Voltage Tolerance	ΔV_{o}	$V_{in} = 12V, I_o = 13.0A$ -40°C $\leq T_a \leq +85$ °C	Vo-0.03	_	Vo+0.03	V
Short-Circuit Threshold	I _{sc}	$V_{in} = 12V$	_	20	30	А
Line Regulation	Reg _{line}	$10.8V \le V_{in} \le 13.2V$, $I_o = 13.0A$	_	±5	_	mV
Load Regulation	Regload	V_{in} = 12V, $0.1 \leq I_o \leq 13.0 A$	_	±10	_	mV
V _o Ripple/Noise	V_n	$V_{in} = 12V, I_o = 13.0A$	_	35	_	mV
Transient Response with C _{out} = 330μF	${f t_{tr}} {f V_{os}}$	I_o step between 6.5A and 13A V_o over/undershoot	_	35 70	_	μSec mV
Efficiency	η	$ \begin{array}{ll} V_{in} = 12V, I_o = 9A & V_o = 5.0V \\ V_o = 3.3V \\ V_o = 2.5V \\ V_o = 1.8V \\ V_o = 1.5V \end{array} $	_	93 90 88 83 81	 	%
Switching Frequency	$f_{ m o}$	$\begin{array}{l} 10.8{\rm V} \le {\rm V_{in}} \le \!$	300	350	400	kHz
Absolute Maximum Operating Temperature Range	Та	—	-40 (2)	—	+85 (3)	°C
Storage Temperature	Ts	_	-40	_	+125	°C
Mechanical Shock		Per Mil-STD-883D, Method 2002.3 1 msec, Half Sine, mounted to a fixture	_	500	_	G's
Mechanical Vibration		Per Mil-STD-883D, Method 2007.2, 20-2000 Hz, Soldered in a PC board	_	15	_	G's
Weight	_	_	_	26	_	grams

(1) The ISR will operate down to no load with reduced specifications. Notes:

(2) For operation below 0°C, Cin and Cout must have stable characteristics. Use either low ESR tantalum or Oscon® capacitors.

(3) See Safe Operating Curves or contact the factory for the appropriate derating.

(4) If the Remote Sense Ground is not used, pin 7 must be connected to pin 8 for optimum output voltage accuracy.

External Capacitors: The PT6725 requires a minimum ouput capacitance of 330μ F with a maximum ESR of $50m\Omega @100$ kHz for proper operation. The maximum allowable output capacitance is $15,000\mu$ F. The input capacitance must be rated for a minimum of 1.6Arms of ripple current. For transient or dynamic load applications, additional capacitance may be required. For more information refer to the application note regarding capacitor selection for this product. Input Filter: An input filter inductor is optional for most applications. The inductor must be sized to handle 6.5ADC with a typical value of 1µH.



13 Amp 12V Input Adjustable Integrated Switching Regulator

Features

- 14A Output Current
- +12V Input
- High Efficiency
- 17-pin Space Saving Package
- Solderable Copper Case
- Adjustable Output Voltage
- Differential Remote Sense
- Short Circuit Protection

Ordering Information -----

PT Series Suffix (PT1234X)

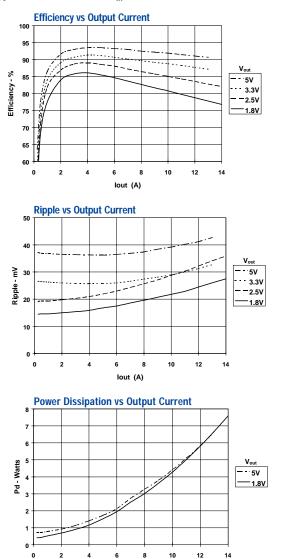
Case/Pin

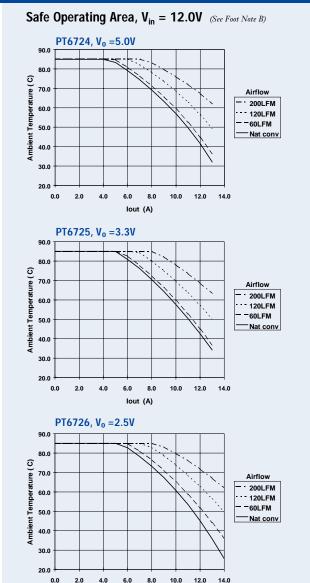
Configuration		
Vertical Through-Hole	Ν	
Horizontal Through-Hole	Α	
Horizontal Surface Mount	С	

(For dimensions and PC board layout, see Package Styles 1340 and 1350.)

TYPICAL C H A R A C T E R I S T I C S

Typical Performance, V_{in} = 12.0V (See Foot Note A)





Note A: All data listed in the above graphs has been developed from actual products tested at 25°C. This data is considered typical data for the ISR. Note B: SOA curves represent the conditions at which internal components are at or below manufacturer's maximum operating temperatures.



lout (A)

6.0 8.0

lout (A)

Capacitor Recommendations for the PT6721 and PT6725 Series Regulators

Input Capacitors

The recommended input capacitance is determined by 1.6 ampere minimum ripple current rating and 1000μ F minimum capacitance. Tantalum capacitors listed below cannot be used on the input bus since they are not rated for 12V operation. Ripple current and Equivalent Series Resistance (ESR) values are the major considerations along with temperature when selecting the proper capacitor.

Output Capacitors

The minimum required output capacitance is 330μ F with a maximum ESR less than or equal to $50m\Omega$. Failure to observe this requirement may lead to regulator instability or oscillation. Electrolytic capacitors have poor ripple performance at frequencies greater than 400kHz, but excellent low frequency transient response. Above the ripple frequency ceramic decoupling capacitors are necessary to improve the transient response and reduce any microprocessor high frequency noise components apparent during higher current excursions. Preferred low ESR type capacitor part numbers are identified in the Table 1 below.

Tantalum Characteristics

Tantalum capacitors are recommended on the output bus but only TPS Series, Sprague 593D/594/595 Series, or Kemet T495/T510 Series. The AVX TPS Series, Sprague Series or Kemet Series tantalum capacitors are recommended over other manufacturer's due to their higher surge current, excellent power dissipation and ripple current ratings. As an example, the TAJ Series by AVX is not recommended. This series exhibits considerably higher ESR, reduced power dissipation and lower ripple current capability. The TAJ Series is a less reliable compared to the TPS series when comparing power dissipation capability.

Capacitor Table

Table 1 identifies the characteristics of capacitors from a number of vendors with acceptable ESR and ripple current (rms) ratings. The suggested minimum quantities per regulator for both the input and output buses are identified.

This is not an extensive capacitor list. The table below is a selection guide for input and output capacitors. Other capacitor vendors are available with comparable RMS ripple current rating and ESR (Equivalent Series Resistance at 100kHz). These critical parameters are necessary to insure both optimum regulator performance and long capacitor life.

Capacitor Vendor/ Series			Capacitor	Characteristics		Qua	ntity	
	Working Voltage	Value(µF)	(ESR) Equivalent Series Resistance	105°C Maximum Ripple Current(Irms)	Physical Size(mm)	Input Bus	Output Bus	Vendor Number
Panasonic FC Series FA Series	35V 25V 25V	680 1000 1000	$\begin{array}{c} 0.043\Omega \\ 0.038\Omega \\ 0.038\Omega \end{array}$	1655mA 1655mA 1690mA	12.5x20 12.5x20 16x15	2 1 1	1 2 1	EEUFC1V681 EEUFC1E102 EEUFC1E102S
United Chemi -con LFVSeries	35V 35V 16V	680 1000 470	$\begin{array}{c} 0.034\Omega \\ 0.038\Omega \\ 0.084\Omega / 2 = 042\Omega \end{array}$	1690mA 1630mA 825mA x2	12.5x25 16x20 10x16	2 1 N/R(1)	1 1 2	LXV35VB680M12X25LL LXV35VB102M16X20LL LXV16VB471M10X16LL
Nichicon PL Series PM Series	35V 25V 35V	680 1200 1000	0.036Ω 0.039Ω 0.034Ω	1660mA 1600mA 1770mA	12.5x25 18x15 16x20	2 1 1	1 1 1	UPL1V681MHH UPL1E122MHH6 UPM1V102MHH6
Panasonic FC Series Surface Mtg	35V 25V 35V	1000 1000 470	0038Ω 0.038Ω 0.0430Ω	2000mA 2000mA 1690mA	18x16.5 18x16.5 16x16.5	1 1 2	1 1 1	EEVFC1V102N EEVFC1E102N EEVFC1V471N
Oscon- SS/SV Series	10V 10V	330 330	0.025Ω 0.020Ω	3500mA 3800mA	10x10.5 10.3x10.3	N/R(1) N/R(1)	1 1	10SS330M(If Output <5V) 10SV330(If Output <5V) Surface Mount(SV)
AVX	10V	330	0.100Ω/2=0.050Ω	>2500mA	7.3L x	N/R(1)	2	TPSE337M010R0100
Tantalum TPS Series	10V	330	0.060Ω/2=0.030Ω	>3000mA	5.7W x 4.1H	N/R(1)	2	TPSV337M010R0060 Surface Mount
Kemet	10V	330	0.033Ω	1400mA	4.3Wx7.3L	N/R(1)	1	510X337M010AS
Tantalum T510/T495 Series	10V	220	0.070Ω/2=0.035Ω	>2000mA	x4.0H	N/R(1)	2	T495X227M0100AS Surface Mount
Sprague Tantalum 594D Series	10V	330	0.045Ω	2360mA	7.2L x 6W x 4.1H	N/R(1)	1	594D337X0010R2T Surface Mount

Table 1 Capacitors Characteristic Data

Note: (N/R) 10V tantalums are not recommend for the input bus.



Adjusting the Output Voltage of the PT6725 Excalibur[™] Converters

The PT6725 ISRs are a non-programmable version of the PT6721, 12V input Excalibur[™] family of converters. These regulators have a fixed output voltage, which may be adjusted higher or lower than the factory trimmed preset voltage using a single external resistor. Table 1 gives the allowable adjustment range for each model as V_a (min) and V_a (max).

Adjust Up: An increase in the output voltage is obtained by adding a resistor R2, between pin 1 (V adjust) and pin 7 (-Remote Sense).

Adjust Down: adjust) and pin 17 (+Remote Sense).

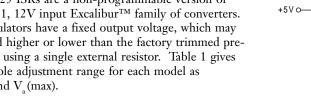
of the required resistor, either (R1) or R2 as appropriate.

Notes:

- 1. Use only a single 1% resistor in either the (R1) or R2 location. Place the resistor as close to the ISR as possible.
- 2. Never connect capacitors from V adjust to either GND, V_{out}, or the Remote Sense pins. Any capacitance added to the V_o adjust pin will affect the stability of the ISR.
- 3. If the remote sense feature is not being used, pin 7 must be connected to pin 8 for optimum output voltage accuracy. Consequently the resistors (R1) and R2 may then be connected from $V_{_{\rm o}}\,Adjust$ to either $V_{_{\rm out}}\,or\;GND$ respectively.

Table 1



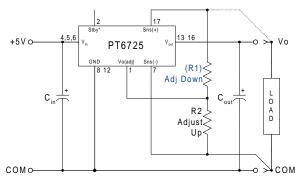


Add a resistor (R1), between pin 1 (V

Refer to Figure 1 and Table 2 for both the placement and value

Figure 1

(



The values of (R1) [adjust down], and R2 [adjust up], can also be calculated using the following formulas.

R1) =
$$\frac{R_{o} \cdot (V_{a} - 1.27)}{(V_{o} - V_{a})} - R_{s} \quad k\Omega$$

$$R2 = \frac{1.27 R_o}{V_a - V_o} - R_s \qquad k\Omega$$

Where: V_0 = Original output voltage

V = Adjusted output voltage

= The resistance value in Table 1 R

R = Series resistance value from Table 1

PT6725 SERIE	S ADJUSTMENT	PARAMETERS				
Series Pt #	PT6728	PT6727	PT6726	PT6725	PT6724	
Vo (nom)	1.5	1.8	2.5	3.3	5.0	
V _a (min)	1.47	1.75	2.25	2.75	4.01	
Va (max)	1.73	2.0	2.85	3.75	5.47	
R ₀ (kΩ)	10.0	10.0	10.0	10.0	9.09	
R _S (kΩ)	49.9	49.9	33.2	24.9	24.9	



Power Trends Products from Texas Instruments

PT6725 Series

Table 2

Series Pt #	PT6728	PT6727	PT6726	PT6725
I₀ (nom)	1.5	1.8	2.5	3.3
I _a (req'd)				
1.47	(16.8)kΩ			
1.5				
1.55	204.0kΩ			
1.6	77.1kΩ			
1.65	34.8kΩ			
1.7	13.6kΩ			
1.75		(46.1)kΩ		
1.8				
1.85		204.0kΩ		
1.9		77.1kΩ		
1.95		34.8kΩ		
2.0		13.6kΩ		
2.05				
2.1				
2.15				
2.2				
2.25			(6.0)kΩ	
2.3			(18.3)kΩ	
2.35			(38.8)kΩ	
2.4			(79.8)kΩ	
2.45			(203.0)kΩ	
2.5			~ /	
2.55			221.0kΩ	
2.6			93.8kΩ	
2.65			51.5kΩ	
2.7			30.3kΩ	
2.75			17.6kΩ	(2.0)k
2.8			9.1kΩ	(5.7)k
2.85			3.1kΩ	(10.2)k
2.9				(15.9)k
2.95				(23.1)k
3.0				(32.8)k
3.05				(46.3)k
3.1				(66.6)k
3.15				(100.0)k
3.2				(168.0)k
3.25				(371.0)k
3.3				(, ,)A
3.35				229.0kg
3.4				102.0kg
3.45				59.8kg
3.5				38.6kg
3.55				25.9kg
3.6				17.4kg
3.65				11.4kg
3.7				6.9kg
3.75				3.3kg

Series Pt #	PT6724
V _o (nom)	5.0
V _a (req'd)	
4.1	(3.7)kΩ
4.2	(8.4)kΩ
4.3	(14.4)kΩ
4.4	(22.5)kΩ
4.5	(33.8)kΩ
4.6	(50.8)kΩ
4.7	(79.0)kΩ
4.8	(136.0)kΩ
4.9	(305.0)kΩ
5.0	
5.1	90.5kΩ
5.2	32.8kΩ
5.3	13.6kΩ
5.4	4.0kΩ

Operating Features of the Programmable PT6700 "Excalibur™" Series ISRs

Power Good

Programmable versions of the PT6700 Series regulators incorporate a *PWR Good* output (pin 2). This output is open-drain and generates an acitve-high signal when the sensed output from the ISR is within a nominal $\pm 10\%$ of the programmed set point. When the regulated output is outside this range, pin 2 asserts a logic low (typically <0.1V). A 10k Ω pull-up resistor to a valid bus voltage is required. If the power good feature is not used, the pull-up resistor can be omitted. The maximum voltage that may be applied to the pull-up resistor is 15V.

Over-Voltage Protection (OVP)

The PT6700 programmable regulators also incorporate an OVP function. The *OVP DRV* (pin 1) normally has a logic low output (typically <0.1V). When the ISR's sensed output exceeds the programmed output setting by 15%, pin 1 produces a 60mA, +12V drive signal. This drive signal can trigger an SCR, which can be used to disable the input voltage, or alternatively interface to another external monitoring device. When the ISR output voltage returns to within 15% of its programmed setting, pin 1 reverts back to its low state. If the OVP function is not used, pin 1 may be left open circuit.

Stand-By Function

The PT6700 series ISRs incorporate a standby function. This feature may be used for power-up sequencing, or wherever there is a requirement for the output voltage to be controlled by external circuitry.

If the *STBY*^{*} input (pin 8) is left open-circuit the regulator operates normally, providing a regulated output when a valid supply voltage is applied to V_{in} (pins 10-12) with respect to GND (pins 14-18). Connecting pin 8 to ground 1 places the regulator in standby mode, and reduces the input current to typically 20mA (30mA max). Applying a ground signal to pin 8 prior to power-up, will disable the output during the period that input power is applied. To ensure that the regulator output is fully enabled, pin 8 must be allowed to rise to a minimum of 3.0V.

Notes:

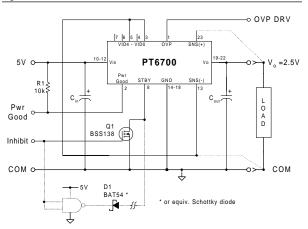
 The standby on the PT6700 series is ideally controlled with an open-collector (or open-drain) discrete transistor (See fig. 1). Table 1 gives the threshold requirements. <u>Do</u> <u>Not</u> use a pull-up resistor. The control input has an opencircuit voltage of about 4.0Vdc. To set the regulator output to zero, the control pin must be "pulled" to less than 1.0Vdc with a sink to ground.

- The Standby input can also be interfaced to TTL or other bi-directional output device using a schottky diode. See Figure 1.
- 3. When placed in the standby mode, the regulator output may assert a low impedance to ground. If an external voltage is applied to the output, it will sink current and possibly over-stress the part.

Table 1 Inhibit Control Threshold 2,3

	Shora	
Min	Тур	Max
3.0V		
-0.1V	0.4V	10V
	0.01mA	
	Min 3.0V	3.0V -0.1V 0.4V

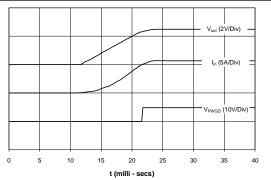
Figure 1



Turn-On Time

Turning Q₁ in Figure 1 off, removes the low-voltage signal at pin 8 and enables the output. Following a brief delay of 10-15ms, the output voltage of the PT6700 regulator rises to full regulation within 30ms. Figure 2 shows the typical output voltage waveform of a PT6701 following the prompt turn-off of Q₁ at time t =0 secs. The output voltage was set to 2.5V. The waveforms were measured with a 5V input source voltage, and 10A resistive load.







Pin-Coded Output Voltage Adjustment on Non-Isolated "Excalibur™" Series ISRs

The PT6701/6702/6721 Excalibur™ ISRs incorporate a pin-coded voltage control to adjust the ouput voltage. The control pins are identified VID0 - VID4 (pins 3–7) respectively. When these control pins are left open-circuit the ISR output will regulate at its factory trimmed output voltage. Each pin is internally connected to a precision resistor, which when grounded changes the output voltage by a set amount. By selectively grounding VID0-VID4, the output voltage these ISRs can be programmed in incremental steps over the specified output voltage range. The program code and output voltage range of these ISRs is compatible with the voltage ID specification defined by Intel Corporation. The code is used in conjuction with voltage regulator modules (VRMs) that are used to power Intel's Pentium® microprocessors. Refer to Figure 1 below for the connection schematic, and the respective device Data Sheet for the appropriate programming code information.

Notes:

1. The programming convention is as follows:-

Logic 0: Connect to pin13 (Remote Sense Ground).

Logic 1: Open circuit/open drain (See notes 2, & 4)

- 2. Do not connect pull-up resistors to the voltage programming pins.
- 3. To minimize output voltage error, always use pin 13 (Remote Sense Ground) as the logic "0" reference. While the regular ground (pins 14-18) can also be used for programming, doing so will degrade the load regulation of the product.

4. If active devices are used to ground the voltage control pins, low-level open drain MOSFET devices should be used over bipolar transistors. The inherent V_{ce}(sat) in bipolar devices introduces errors in the device's internal voltage control circuit. Discrete transistors such as the BSS138, 2N7002, IRLML2402, or the 74C906 hex open-drain buffer are examples of appropriate devices.

Active Voltage Programming:

Special precautions should be taken when making changes to the voltage control progam code while the unit is powered. It is highly recommended that the ISR be either powered down or held in standby. Changes made to the program code while Vout is enabled induces high current transients through the device. This is the result of the electrolytic output capacitors being either charged or discharged to the new output voltage set-point. The transient current can be minimized by making only incremental changes to the binary code, i.e. one LSB at a time. A minimum of 100µs settling time between each program state is also recommended. Making non-incremental changes to VID3 and VID4 with the output enabled is discouraged. If they are changed, the transients induced can overstress the device resulting in a permanent drop in efficiency. If the use of active devices prevents the program code being asserted prior to power-up, pull pin 8 (STBY) to the device GND during the period that the input voltage is applied to V_{in} . Releasing pin 8 will then allow the device output to initiate a soft-start power-up to the programmed voltage.

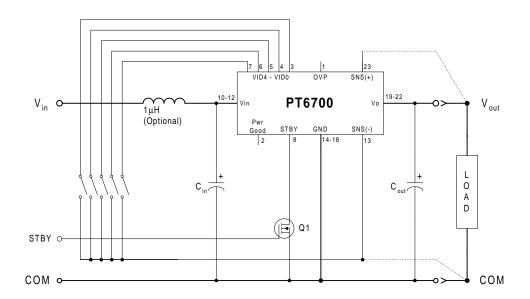


Figure 1

